RICHTEK®

RT4531

Dual Channel 10s2p WLED Drivers

General Description

The RT4531 is a dual-channel current source with high-efficiency asynchronous boost converter for backlight application. The RT4531 has built-in 1.5A/36V power MOSFET and two high current-matching capability current sink, so RT4531 can drives up 10s2p white LED. The RT4531 can allow very low voltage headroom control, therefore to improve LED strings efficiency effectively.

The RT4531 supports both PWM dimming and 1-wire digital dimming interface and it can realize 10 bit brightness code programming.

The RT4531 provides complete protection functions such as input Under-Voltage Lockout (UVLO), Over-Current Protection (OCP), Output Over Voltage (OVP) and Over-Temperature Protection (OTP).

The RT4531 is available in the WL-CSP-9B 1.35x1.35 (BSC) package.

Applications

- Smart Phones
- Probable Instruments
- Backlight for Small and Media Form
- Factor LCD Display with Single-cell Battery Input

Features

- **Input Voltage Range : 2.5V to 5.5V**
- **Internal Soft-Start, UVLO, OTP, OCP, OVP**
- **Typical Shutdown Current : < 1μA**
- **1MHz Switching Frequency**
- **Drives Up to 10 WLEDs in Two Strings**
- **Independent PWM Dimming and 1-Wire Dimming Control**
- **PWM Dimming Frequency from 20k to 100kHz**
- **Up to 10 Bit Dimming Resolution**
- **LED Current Accuracy ± 2% (I_{FBx} = 20mA)**
- **LED Current Matching ± 2% (I_{FBx} = 20mA)**
- **Built-in IFB1/IFB2 pin OVP, Short Protection, and Un-Use Detection**

## Ordering Information

RT4531☐

    └─Package Type
      WSC : WL-CSP-9B 1.35x1.35 (BSC)

Note :

Richtek products are :

▸ RoHS compliant and compatible with the current require-ments of IPC/JEDEC J-STD-020.

▸ Suitable for use in SnPb or Pb-free soldering processes.

Simplified Application Circuit

RT4531
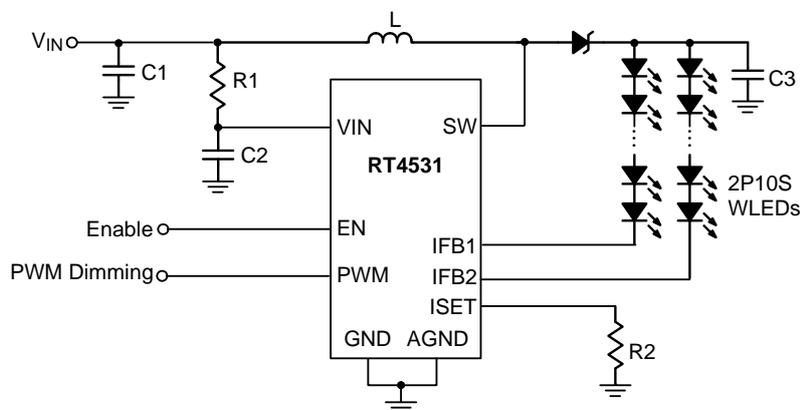
Pin Configurations

(TOP VIEW)

```
                IFB2
                 |
ISET  (A1) (A2) (A3)  IFB1
PWM   (B1) (B2) (B3)  GND
           AGND
EN    (C1) (C2) (C3)  SW
                 |
                VIN
```
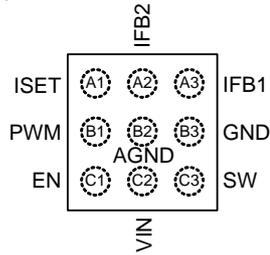
WL-CSP-9B 1.35x1.35 (BSC)

Marking Information

```
┌─────────┐
│         │
│   33W   │
│         │
● └─────────┘
```
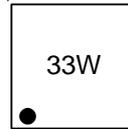
33 : Product Code

W : Date Code

Functional Pin Description

Pin No.	Pin Name	Pin Function
A1	ISET	Full-Scale LED Current Set Pin. Connecting a resistor to the pin programs the full-scale LED current.
A2	IFB2	Single Output 2 for Backlight LED.
A3	IFB1	Single Output 1 for Backlight LED.
B1	PWM	Enable Control, and PWM Dimming Signal Input.
B2	AGND	Analog Ground.
B3	GND	Power Ground.
C1	EN	Enable Control, and 1-Wire Dimming Signal Input.
C2	VIN	Supply Input Pin.
C3	SW	Switch Node of Boost Converter.

Function Block Diagram

Operation

The RT4531 is a high efficiency solution with 10S2P string for backlight applications. The RT4531 optimizes the feedback regulation voltage to provide high efficiency and also keep high current matching. The RT4531 can receive either the 1-wire dimming at EN pin or PWM dimming at PWM pin for brightness dimming. If the 1-wire dimming is selected, the PWM pin should be kept high; If the PWM dimming is selected, the EN pin should be kept high.

OCP Protection

The RT4531 features a 1.5A OCP. The driver provides cycle-by-cycle current limit function to control the current on power switch. The boost switch turns off when the inductor current reaches this current threshold and it remains off until the beginning of the next switching cycle. This protects the RT4531 and external component under overload conditions.

OTP Protection

The OTP function will disable boost and current source when the junction temperature exceeds 150°C. Then restart to soft-start when the junction temperature is smaller than 135°C.

OVP Protection

The OVP function monitors the SW pin's voltage. Once it exceeds OVP threshold = 36.5V twice, the boost and current source will be latched off. Until PWM or EN go high from low again, the latch state will be released.

1-wire Brightness Dimming

The RT4531 is built-in a 10-bit resolution brightness control. The EN pin features a simple 1-wire interface to allow digital brightness control. If the 1-wire dimming is needed, signals of a specific pattern should be input detection window into EN pin; Otherwise, PWM pin need to be enabled.(refer to Timing Diagram1 for Continuous Coding).The Register Program and register map give an overview of the protocol used by RT4531.

PWM Brightness Dimming

Besides programmable built-in 1-wire backlight LED current control, the RT4531 features a built-in PWM dimming current control by PWM pin, offering a linear current dimming by external clock source. In order to guarantee the PWM dimming resolution, recommending PWM dimming frequency have to be operated at range of 20kHz to 100kHz.

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Absolute Maximum Ratings   (Note 1)

- Supply Voltage, V_{IN} --- −0.3V to 6V
- LED Switching Voltage, SW --- 38V
- LED Current Source Voltage, IFB1, IFB2 -- −0.3V to 6V
- Digital Clock Control Pin, EN, PWM --- −0.3V to 6V
- Other Pins, ISET --- −0.3V to 6V
- Power Dissipation, P_D @ T_A = 25°C
  WL-CSP-9B 1.35x1.35 (BSC) -- 1.22W
- Package Thermal Resistance (Note 2)
  WL-CSP-9B 1.35x1.35 (BSC), θ_{JA} -- 81.5°C/W
- Junction Temperature -- 150°C
- Lead Temperature (Soldering, 10 sec.) -- 260°C
- Storage Temperature Range -- −65°C to 150°C
- ESD Susceptibility   (Note 3)
  HBM (Human Body Model) --- 2kV
  MM (Machine Model) --- 200V

Recommended Operating Conditions   (Note 4)

- Supply Input Voltage, V_{IN} -- 2.5V to 5.5V
- Junction Temperature Range --- −40°C to 125°C
- Ambient Temperature Range --- −40°C to 85°C

Electrical Characteristics

(V_{IN} = 3.6V, C_{IN} = 2.2µF, C_{OUT} = 1µF, L = 4.7µH, T_A = 25°C, unless otherwise specified)

| Parameter	Symbol	Test Conditions	Min	Typ	Max	Unit
Power Supply						
UVLO	V_{UVLO}	VIN falls	--	2.2	2.3	V
UVLO Hysteresis Voltage	ΔV_{UVLO}	VIN rises after UVLO	30	50	200	mV
VIN Supply Current	I_Q	EN = High, Non-switching	--	1.2	--	mA
VIN Shutdown Current	I_{SHDN}	V_{IN} = 5V	--	1	--	µA
Thermal Shutdown	T_{SD}		--	150	--	°C
Thermal Shutdown Hysteresis			--	15	--	°C
Back Light LED						
Accuracy of Back Light Output Current 1/2	AIBL1, 2	I_{FBx} = 20mA	−2	--	2	%
Matching of Output Current	MIBL	I_{FBx} = 20mA	--	1	2	%
		I_{FBx} = 500µA	--	1	--	%
BLED Sense Voltage 1/2	VDSBL	I_{FBx} = 20mA, measured on IFBx pin which has a lower voltage.	200	300	430	mV
Back Light Switching Frequency	f_{blosc}		0.88	1	1.12	MHz

Parameter		Symbol	Test Conditions	Min	Typ	Max	Unit
Maximum Duty Cycle		D_{MAX}		90	95	--	%
Back Light N-MOSFET Ron		$R_{DS(ON)\_BLN}$		100	270	400	mΩ
Back Light Current Limit		I_{LIM_BL}	After 5ms in soft-start	1.15	1.5	1.85	A
Back Light Over Voltage Protection		V_{BLOVP}	Close loop, V_{IN} = 3.6V	35.5	36.5	37.5	V
IFBx Pin Over Voltage Threshold		$V_{IFBXOVP}$	Measure on IFBx pin	--	5	--	V
Logic Control							
One Wire Detection Delay		tes_delay	Measure from En low to high	300	--	--	µs
One Wire Detection Time		tes_det	En pin low time in detection window	400	--	--	µs
One Wire Detection Window		tes_win	Measure from En low to high, Fblosc = 1MHz	--	1	--	ms
Start Time of Program Stream		t_{start}		3	--	--	µs
End Time of Program Stream		t_{EOS}		3	--	--	µs
High Time Low Bit		t_{H_LB}	Logic 0	2	--	180	µs
Low Time Low Bit		t_{L_LB}	Logic 0	2 x tH_LB	--	360	µs
High Time High Bit		t_{H_HB}	Logic 1	2 x tL_HB	--	360	µs
Low Time High Bit		t_{L_HB}	Logic 1	2	--	180	µs
EN Input Voltage	Logic-Low	V_{EN_L}		--	--	0.4	V
	Logic-High	V_{EN_H}		1.4	--	--	
PWM Input Voltage	Logic-Low	V_{PWM_L}		--	--	0.4	V
	Logic-High	V_{PWM_H}		1.4	--	--	
EN Pull Low Resistor		R_{EN}		280	400	520	kΩ
PWM Pull Low Resistor		R_{PWM}		280	400	520	kΩ
PWM Logic Low Width to Shutdown		T_{SD_PWM}	PWM pin high to low, then shutdown	20	--	--	ms
EN Logic Low Width to Shutdown		T_{DS_EN}	EN pin high to low, then shutdown	2.5	--	--	ms

Note 1. Stresses beyond those listed ″Absolute Maximum Ratings″ may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions may affect device reliability.

Note 2. θ$_{JA}$ is measured at T_A = 25°C on a high effective thermal conductivity four-layer test board per JEDEC 51-7.

Note 3. Devices are ESD sensitive. Handling precaution is recommended.

Note 4. The device is not guaranteed to function outside its operating conditions.

## Typical Application Circuit
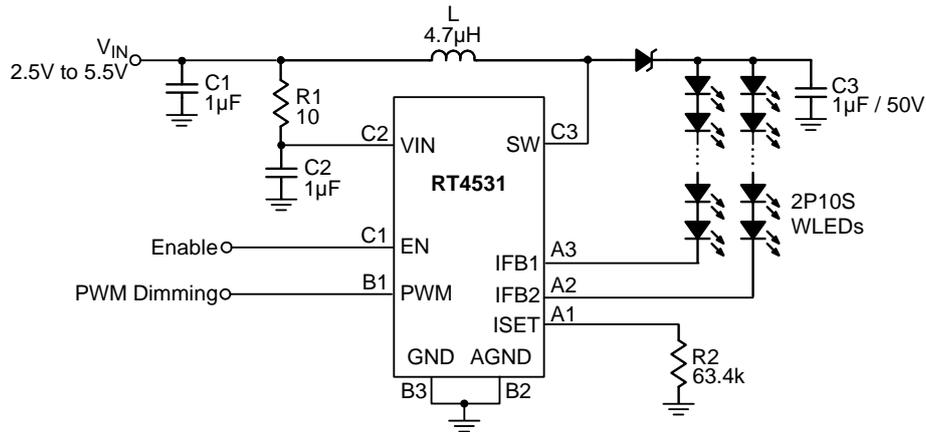
## Timing Diagram

### Timing Diagram1 for 1-wire Continuous Coding
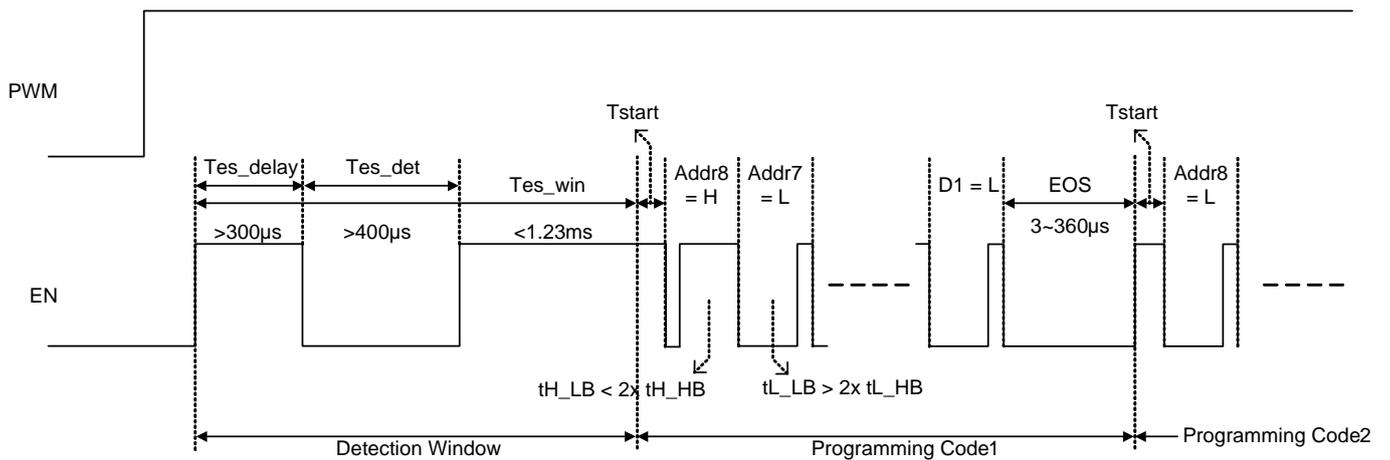

Timing Diagram2 for Shutdown and then Reprogram
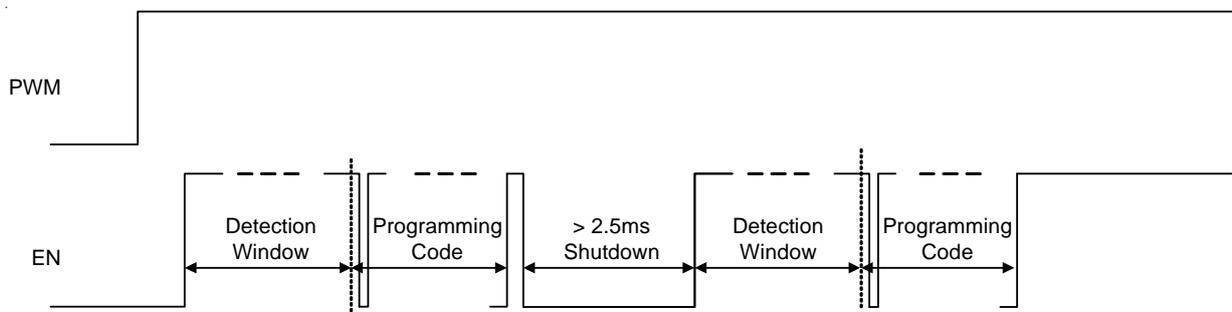
Typical Operating Characteristics

Efficiency vs. Output Current

V_{IN} = 3.6V, V_{OUT} = 31V, 10s2p, 20mA/string
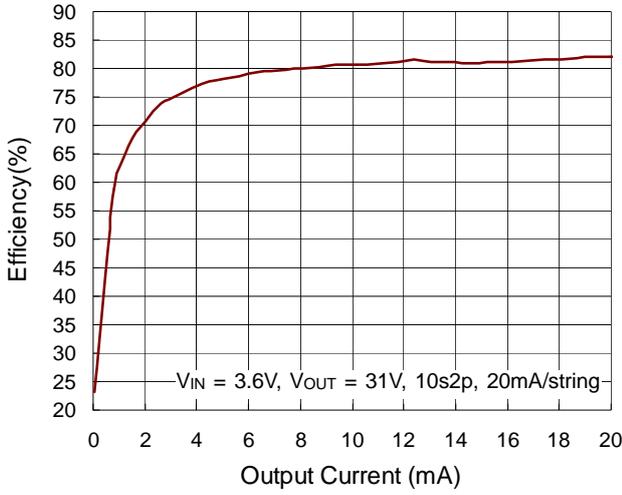
Efficiency vs. Input Voltage

I_{LED} = 20mA
I_{LED} = 10mA

V_{OUT} = 31V, 10s2p
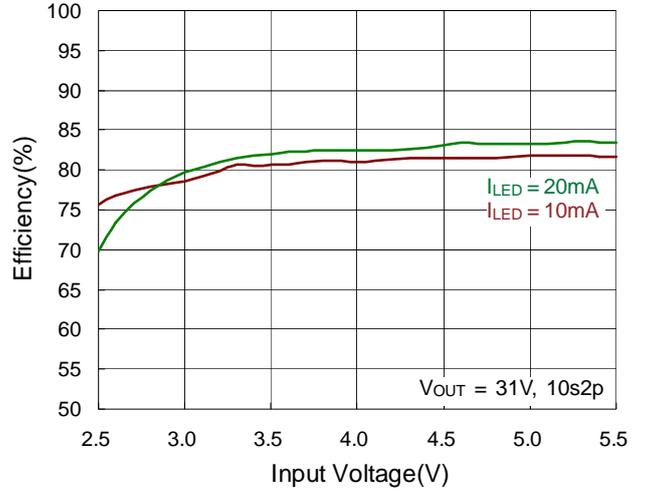
Dimming Linearity

I_{LED1}
I_{LED2}

V_{IN} = 3.6V
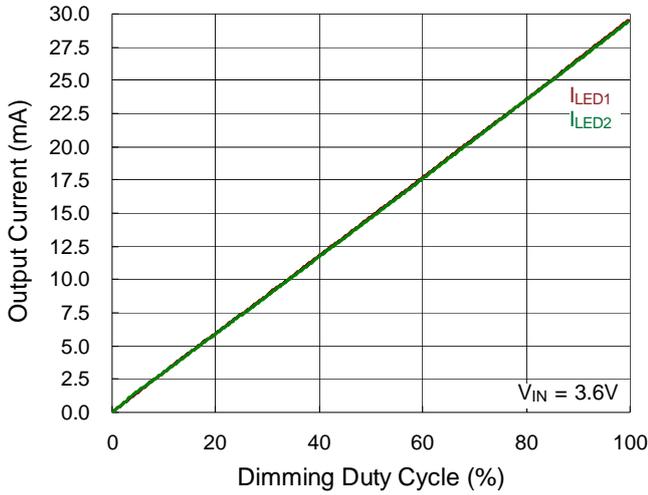
Switching Frequency vs. Input Voltage
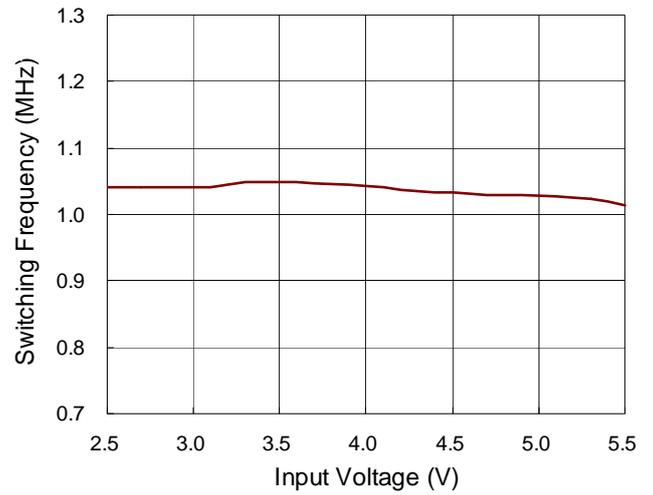
Switching Frequency vs. Temperature
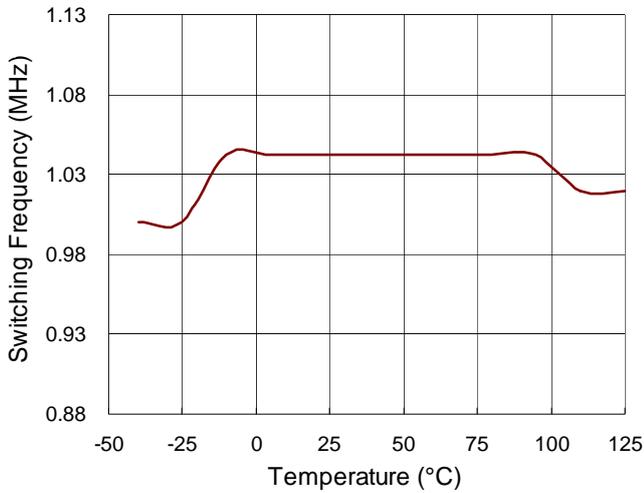
Sense Voltage vs. Output Current

RT4531
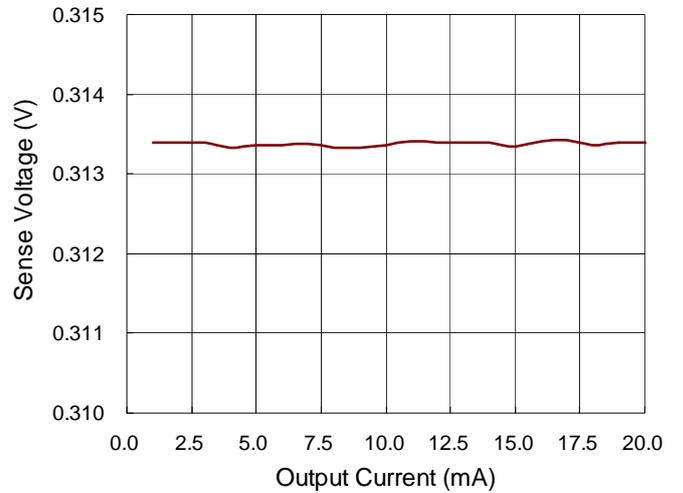
Application Information

Register Program

1-Wire Interface (Device Address 0x2F)

Start	Addr 8	Addr 7	Addr 6	Addr 5	Addr 4	Addr 3	Addr 2	Addr 1	D16	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	EOS

Address — Data3 — Data2 — Data1
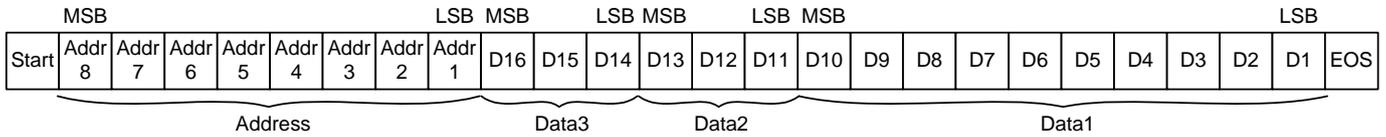
Register Map

Bit	Label	Reset	Description
D16	Ramp Up<2>	0	[000] = 4µs, [001] = 8µs, [010] = 16µs,[011] = 64µs, [100] = 256µs, [101] = 1.024ms, [110] = 2.048ms, [111] = 4.096ms
D15	Ramp Up<1>	0	
D14	Ramp Up<0>	1	
D13	Ramp Down<2>	0	[000] = 4µs, [001] = 8µs, [010] = 16µs, [011] = 64µs, [100] = 256µs, [101] = 1.024ms, [110] = 2.048ms, [111] = 4.096ms
D12	Ramp Down<1>	0	
D11	Ramp Down<0>	1	
D10	ILED<9>	0	MSB = sDAC<9>
D9	ILED<8>	0	sDAC<8>
D8	ILED<7>	0	sDAC<7>
D7	ILED<6>	0	sDAC<6>
D6	ILED<5>	0	sDAC<5>
D5	ILED<4>	0	sDAC<4>
D4	ILED<3>	0	sDAC<3>
D3	ILED<2>	0	sDAC<2>
D2	ILED<1>	0	sDAC<1>
D1	ILED<0>	0	LSB = sDAC<0>

Protection Actions
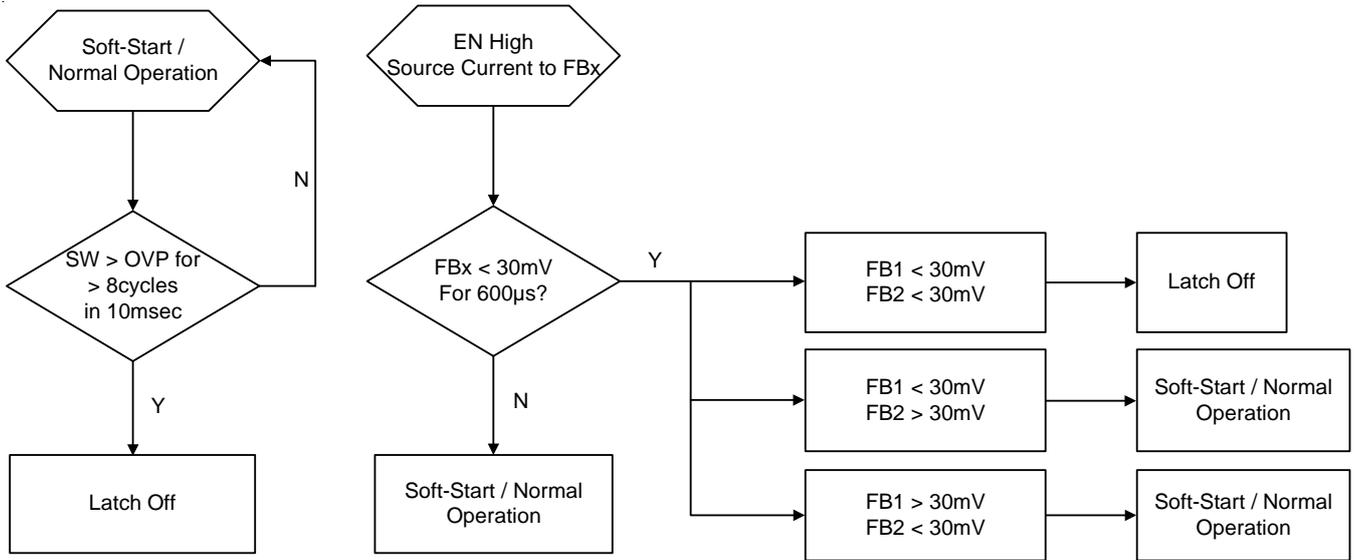
LED Current Setting

The dual channels of the RT4531 can provide up to 30mA current each.

No matter either 1-wire Brightness Dimming or PWM Brightness Dimming is selected, the full-scale current (current when dimming duty cycle is 100%) of each channel should be programmed by an external resistor RISET at ISET pin according to Equation 1.

$$I_{FB} = \frac{V_{ISET}}{R_{ISET}} \times K_{ISET} \qquad (1)$$

Where :

V_{ISET} = 1V,

K_{ISET} = 1268

Brightness Control

a. 1-wire Brightness Dimming

The RT4531 is built-in a 10-bit resolution brightness control. The EN pin features a simple 1-wire interface to allow digital brightness control. If the 1-wire dimming is needed, signals of a specific pattern should be input detection window into EN pin; Otherwise, PWM pin need to be enabled.(refer to Timing Diagram1 for Continuous Coding).The Register Program and register map give an overview of the protocol used by RT4531.By the 1-wire Brightness Dimming, a master can program the 10-bit code D1(LSB) to D10(MSB) to any of 1024 steps with a single command. The programmed value will be stored in an internal register and set the dual-channel current according to Equation 2. The code will be reset to default value when the IC is shut down or disabled.

$$I_{FBX} = I_{FB} \times \frac{DAC\ code}{1024} \qquad (2)$$

Where :

I_{FB} : the full-scale LED current set by the RISET at ISET pin.

Code : the 10-bit DAC code D1~D10 programmed by 1-wire interface.

To enter 1-wire dimming mode, the following digital pattern on the EN pin must be recognized by the IC when the IC starts from the shutdown mode.

1. Pull the EN pin high to enable the RT4531 and start the detection window for digital dimming.

2. After the digital dimming detection delay time (T_es_delay, >300μs), drive the EN low for more than the detection time (T_es_det, >400μs).

3. Pull the EN pin high after the detection time (>400μs) and before the detection window (Tes_win, <1.23ms), once the above 3 conditions are met, the IC immediately enters the 1-wire dimming mode. The digital dimming communication can start before the detection window expires. Once the dimming mode is selected, it cannot be changed without another start up. This means the IC needs to be shut down by pulling the EN low for 2.5ms and restarts. See the dimming mode detection and soft-start for a graphical explanation. (see Timing Diagram1 for 1-wire Continuous Coding).
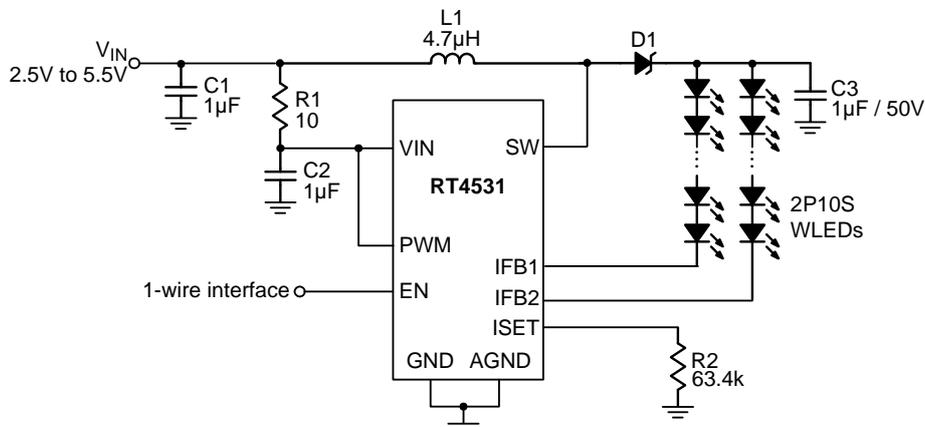

Figure 1. RT4531 1-wire Dimming Typical Application

b. PWM Brightness Dimming

Besides programmable built-in 1-wire backlight LED current control, the RT4531 features a built-in PWM dimming current control by PWM pin, offering a linear current dimming by external clock source. In order to guarantee the PWM dimming resolution, recommending PWM dimming frequency have to be operated at range of 20kHz to 100kHz. The PWM mode ramp time is 64μs each step.

When the PWM pin is constantly high, the dual channel current is regulated to full-scale according to Equation 1.

The PWM pin allows PWM signals to reduce this regulation current according to the PWM duty cycle; therefore,

it achieves LED brightness dimming. The relationship between the PWM duty cycle and I_{FBx} current is given by Equation 3.

$$I_{FBX} = I_{FB} \times Duty\,(\%) \tag{3}$$

Where :

I_{FBX} is the current of each current sink, I_{FB} is the full-scale LED current, Duty is the duty cycle information detected from the PWM signals.
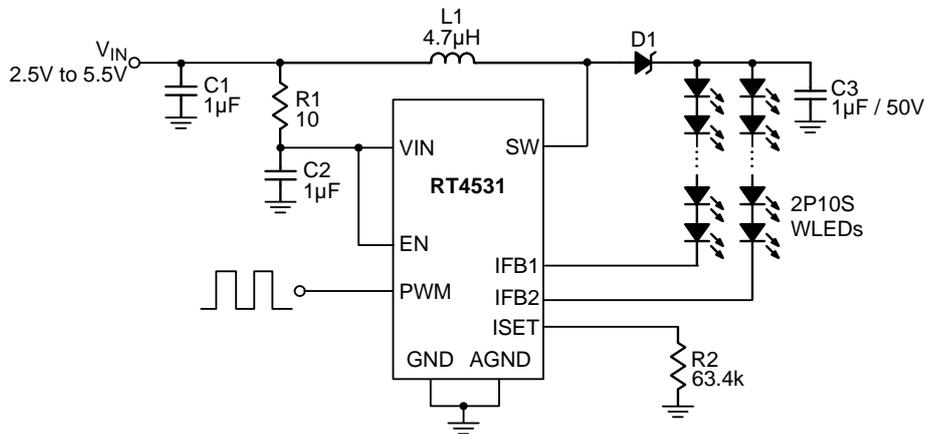

Figure 2. RT4531 PWM Dimming Typical Application

OCP Protection

The RT4531 features a 1.5A OCP. The driver provides cycle-by-cycle current limit function to control the current on power switch. The boost switch turns off when the inductor current reaches this current threshold and it remains off until the beginning of the next switching cycle. This protects the RT4531 and external component under overload conditions.

OTP Protection

The OTP function will disable boost and current source when the junction temperature exceeds 150°C. Then restart to soft-start when the junction temperature is smaller than 135°C.

Open LED Protection

Open LED protection circuitry prevents IC damage as the result of LED disconnection. The RT4531 monitors the voltage at the LX pin during each switching cycle. The circuitry turns off the switch and shuts down the IC as soon as the LX voltage exceeds the VOVP threshold. The device remains in shutdown mode until it is enabled by toggling the EN pin logic. The RT4531 is designed to work with inductor values for 4.7μH (Taiyo NR4018T4R7M) and Cout values for 1μF/50V (Murata 0603/X5R)that the OVP is 37.5V typ. Generally, the VF of the backlight LED is about 3.1 to 3.3V, the RT4531 can support up to 10 LEDs and it would not be trigger OVP.

Thermal Considerations

For continuous operation, do not exceed absolute maximum operation junction temperature. The maximum power dissipation depends on the thermal resistance of IC package, PCB layout, the rate of surroundings airflow and temperature difference between junction to ambient. The maximum power dissipation can be calculated by following formula :

$$P_{D(MAX)} = (T_{J(MAX)} - T_A) / \theta_{JA}$$

Where $T_{J(MAX)}$ is the maximum operation junction temperature, T_A is the ambient temperature and the θ_{JA} is the junction to ambient thermal resistance.

For recommended operating condition specifications, the maximum junction temperature is 125°C. The junction to ambient thermal resistance, θ_{JA}, is layout dependent. For WL-CSP-9B 1.35x1.35 (BSC) package, the thermal resistance, θ_{JA}, is 81.5°C/W on the standard JEDEC 51-7 four-layer thermal test board. The maximum power dissipation at T_A = 25°C can be calculated by following formula :

$$P_{D(MAX)} = (125°C - 25°C) / (81.5°C/W) = 1.22W \text{ for}$$
WL-CSP-9B 1.35x1.35 (BSC) package

The maximum power dissipation depends on operating ambient temperature for fixed $T_{J(MAX)}$ and thermal resistance θ_{JA}. The derating curve in Figure 3 of derating curves allows the designer to see the effect of rising ambient temperature on the maximum power allowed.

Layout Consideration

For best performance of the RT4531, the following guidelines must be strictly followed.

‣ Input and Output capacitors should be placed close to the IC and connected to ground plane to reduce noise coupling.

‣ The GND and Exposed Pad should be connected to a strong ground plane for heat sinking and noise protection.

‣ Keep the main current traces as possible as short and wide.

‣ SW node of DC/DC converter is with high frequency voltage swing. It should be kept at a small area.

‣ It recommended to separate the ISET line and PWM line that avoid the PWM signal couple to ISET.
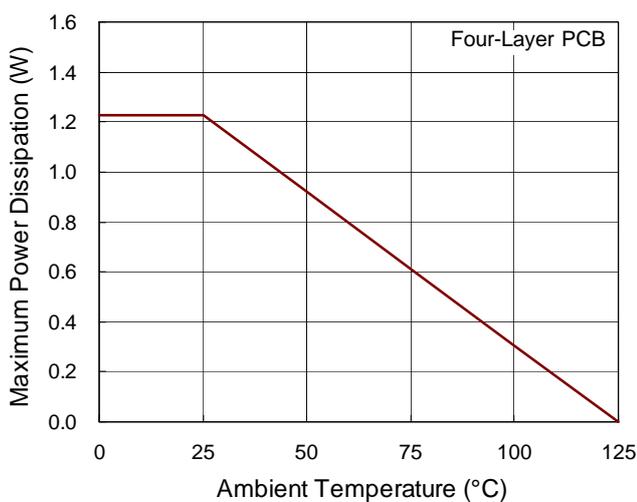
Figure 3. Derating Curve of Maximum Power Dissipation

Cout must be placed between GND and VOUT to reduce noise.

It recommended to separate the ISET line and PWM line that avoid the PWM signal couple to ISET.

CIN should be placed as closed as possible to VIN pin for good filtering.

The inductor should be placed as close as possible to the switch pin to minimize the noise coupling into other circuits. SW node copper area should be minimized for reducing EMI.

Figure 4. PCB Layout Guide

Table 1. Recommended Components for Typical Application Circuit

Reference	Part Number	Description	Manufacture
Schottky Diode	SS0540	Schottky Diode	PANJIT
C1	EMK107BJ105MA-T	Capacitor, Ceramic, 1µF/16V X5R	Taiyo Yuden
R1	RC0603FR-0763K4	Resistor 63.4kΩ, 1%	YAGEO
C2	GRM188R61H105KAAL	MLCC,0603,X5R,50V,1uF	Murata
L	NR4018T4R7M	Inductor, 4.7µH	Taiyo Yuden
C3	EMK107BJ105MA-T	Capacitor, Ceramic, 1µF/16V X5R	Taiyo Yuden
R2	RC0603FR-0710R	Resistor 10Ω, 1%	YAGEO

Outline Dimension
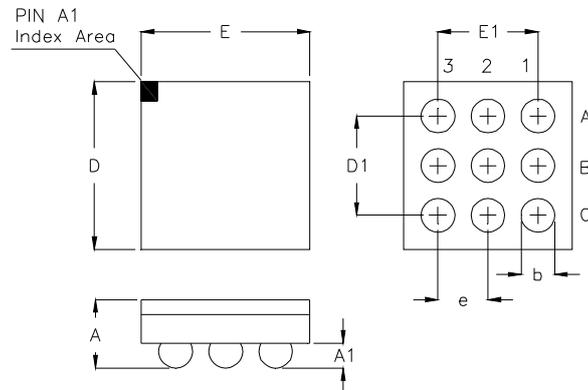
Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min.	**Max.**	**Min.**	**Max.**
A	0.500	0.600	0.020	0.024
A1	0.170	0.230	0.007	0.009
b	0.240	0.300	0.009	0.012
D	1.300	1.400	0.051	0.055
D1	0.800		0.031	
E	1.300	1.400	0.051	0.055
E1	0.800		0.031	
e	0.400		0.016	

WL-CSP-9B 1.35x1.35 (BSC)

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